## EECS 388: Computer Systems and Assembly Language Homework 5 Solution

1. (20) How many RTI interrupt events must occur to generate a 15 minute delay assuming the MCLK is operating at 2 MHz and the RTR[2:0] bits are set for " 110 "? How do you set up the Real_Time Interrupt Control Register (RTICTL) (i.e., enable RTI and set RTI pre-scale) for this purpose?

According to the table on page 229 of your textbook, the period of a RTI interrupt is set by the MCLK frequency divided by a divisor stored in RTR[2:0]. Therefore:

- $\quad$ RTR[2:0] $=$ " 110 " implies a clock divider of $2^{18}$.
- $\quad \mathrm{MCLK}=2 \mathrm{MHz}$ and a divider of $2^{18}$ implies that the frequency of RTI interrupts is $7.63 \mathrm{~Hz}\left(2 \mathrm{MHz} / 2^{18}\right)$.
- The time delay, or period, between RTI events is thus $1 / 7.63 \mathrm{~Hz}=0.13$ seconds (because period $=1 /$ frequency).

Now, if we want a delay of 15 minutes:

- 15 minutes $=900$ seconds.
- Number of RTI events for 15 minutes =
- $(900$ seconds $) /(0.13$ seconds per delay $)=6,293.08$.
- Thus, it will require about 6,294 RTI events

One must write the appropriate values into the RTICTL in order to enable interrupts and to set the RTI frequency. The RTICTL register is a memorymapped location located at address $\$ 0014$. The RTIE bit is the MSB (bit 7), while the RTR[2:0] bits are the least-significant 3 bits (bits 2 through 0 ).

| RTICTL | EQU \$0014 $;$ Equate for the address of the RTICTL register |  |
| :--- | :--- | :--- |
|  | LDAA \%10000110 | ; RTICTL mask (RTIE ='1', RTR[2:0] = " $110 "$ ) |
|  | STAA RTICTL | ; Store the value into RTICTL |

2. (15) Textbook, page 291. Advanced problem \#4. Change MCLK to 4 MHz.
Assuming MCLK is at 4 MHz (as stated above), and the pre-scaler is set to 1 , this means that the timing frequency is $(4 \mathrm{MHz}) /\left(2^{1}\right)=2 \mathrm{MHz}$, or a period of 500 ns .

If the two counts (or timestamps) are $\$ 1993$ and $\$ 0 C 78$, then the period of the measured signal (assuming no counter rollovers) is \$D1B, which is 3,355 counter ticks in decimal. This period, in real-time, is thus:

[^0]
## 3. (15) Textbook, page 291. Advanced problem \#5.

If the period of the pulse being measured is greater than the rollover time fo the counter, then one must make sure to detect counter rollovers in order to accurately measure the signal of interest. Think of this process as noting every New Years Eve from when you were born until the present time in order to figure out how old you are. This requires one to modify the program to log every counter rollover (pulse-accumulator overflow bit, or PAOVF).

Given the numbers from above (problem \#2), we know that counter is adjusted by 1 every 500 ns , and that the counter will rollover when it reaches $2^{16}$, or 65,536 . This means that pulse length of interest can be found by counting counter rollovers:

$$
\begin{array}{|l}
\hline \text { Period= \# rollovers + \# of extra ticks } \\
=(500 \mathrm{~ns} / 1 \text { tick) *[(\# of rollovers)*(65,536 ticks / } 1 \text { rollover)+ (current ticks)] }
\end{array}
$$

4. (25) Write a program to measure the period of a periodic signal connected to input channel 3 by measuring the count difference between two falling edges. Set PR2:PR0 $=011$. Use polling method.

This program is very much like the example program found on pg .273 of the textbook.


5. (25) Generate a 1500 Hz square wave with a $40 \%$ duty cycle (ON/PERIOD) on output compare channel 2 (OC2). $\mathrm{MCLK}=8 \mathrm{MHz}$. Set the pre-scaler to divide by 4 . Use interrupt.

This program is very much like the example program found on pg. 275 of the textbook. If the MCLK runs at 8 MHz and the pre-scaler is set to 4 , then the counter will adjust at a rate of $(8 \mathrm{MHz}) / 4=2 \mathrm{Mhz}$, or with a period of 500 ns .

A 1500 Hz signal has a period of 666.67 microseconds, and a $40 \%$ duty cycle means that it will be high for $0.4^{*} 666.67$ microseconds or 266.67 microseconds, and low for 400 microseconds. This translates to counter value of:

High counter:
$=266.67$ microseconds * (1 tick / 0.5 microseconds)
$=534$ ticks $\rightarrow \$ 0216$
Low counter:

$$
=400 \text { microseconds * (1 tick / } 0.5 \text { microseconds) }
$$

$$
=800 \text { ticks } \rightarrow \$ 0320
$$

The program on p. 275 can be modified to perform the $40 \%$ duty cycle switch by changing the following:

- TMSK2_IN needs to be changed to $\$ 04$
- This sets the correct pre-scale for our problem.
- Pre-scale of $4\left(2^{2}=4\right)$.
- The first instance of \#\$03E8 (in TIMERINIT) needs to be changed to the counter value for our high counter, or \#\$0216.
- The instance of \#\$03E8 (in SQWAVE) needs to be changed to alternate between the high and low-counter values using a conditional branch to implement an IF statement.
- This can be easily done by reserving a word (double-byte) in memory to contain the count value. This value will "flip-flop" between the high and low counter value during every iteration of the loop (see the following modifications.

```
.**************************
; New data section
*************************
ORG $6000
COUNT_INC FDB $0000 ; Location to hold counter increment
*
<Within TIMERINIT>
    ; Initialize COUNT_INC to be low counter value
    LDD $0320
    STD COUNT_INC
<Replace SQWAVE with the following>
SQWAVE
    BRCLR TFLG1,$04,SQWAVE ; Poll for counter flag
    LDD TC2H ; Load in counter value
    LDX COUNT_INC ; Load in current COUNT_INC
    CPX #$0216
    ; Compare to high-count
    BEQ ADD_LOW ; If high }->\mathrm{ add low
    ADDD #$0216 ; Otherwise, add high
    LDX #$0216 ; Update COUNT_INC
    STX COUNT_INC
    BRA ENDIF
ADD_LOW
    ADDD #$0320 ; Add low
    LDX #$0320 ; Update COUNT_INC
    STX COUNT_INC
ENDIF
    STD TC2H ; Setup next transition time
    JSR CLEARFLAG ; generate repetitive signal
    RTS ; return
```


[^0]:    3,355 ticks * $500 \mathrm{~ns} / 1$ tick $)=1677500 \mathrm{~ns} \rightarrow 1.6775 \mathrm{~ms}$.

